

Ultralow Power Single/Dual Comparator with Reference

FEATURES

- Ultralow Quiescent Current: 2.1µA Typ (LTC1440)
- Wide Supply Range: Single: 2V to 11V Dual: ±1V to ±5.5V
- Input Voltage Range Includes the Negative Supply
- Reference Output Drives 0.01µF Capacitor
- Adjustable Hysteresis (LTC1440/LTC1442)
- TTL/CMOS Compatible Outputs
- 12µs Propagation Delay with 10mV Overdrive
- No Crowbar Current
- 40mA Continuous Source Current
- Pin Compatible Upgrades for MAX921/922/923

APPLICATIONS

- Battery-Powered System Monitoring
- Threshold Detectors
- Window Comparators
- Oscillator Circuits

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DESCRIPTION

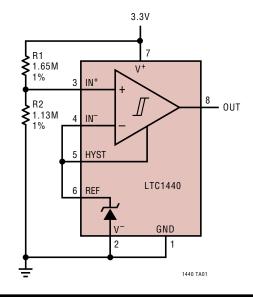
The LTC®1440/LTC1441/LTC1442 are ultralow power single and dual comparators with built-in references. The comparators feature less than $3.7\mu A$ supply current over temperature (LTC1440), a $1.182V~\pm1\%$ reference, programmable hysteresis (LTC1440/LTC1442) and TTL/CMOS outputs that sink and source current. The reference output can drive a bypass capacitor of up to $0.01\mu F$ without oscillation.

The comparators operate from a single 2V to 11V supply or a dual \pm 1V to \pm 5.5V supply (LTC1440). Comparator hysteresis is easily programmed by using two resistors and the HYST pin (LTC1440/LTC1442). Each comparator's input operates from the negative supply to within 1.3V of the positive supply. The comparator output stage can continuously source up to 40mA. By eliminating the cross-conducting current that normally happens when the comparator changes logic states, the power supply glitches are eliminated.

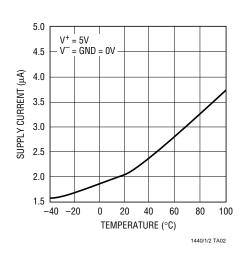
The LTC1440 is available in 8-pin PDIP, SO and MSOP packages. The LTC1441/LTC1442 are available in 8-pin PDIP and SO packages.

TYPICAL APPLICATION

Micropower 2.9V V_{CC} Threshold Detector



LTC1440 Supply Current vs Temperature





ABSOLUTE MAXIMUM RATINGS

Voltage	
V+ to V-, V+ to GND, GND to V12V	to -0.3V
IN^{+} , IN^{-} , HYST($V^{+} + 0.3V$) to (V	-0.3V)
REF(V+ + 0.3V) to (V	-0.3V)
OUT (LTC1440) (V + 0.3V) to (GNI	0.3V
OUT (LTC1441/LTC1442) (V+ + 0.3V) to (V	-0.3V)
Current	
IN+, IN-, HYST	20mA
REF	20mA
OUT	50mA

OUT Short-Circuit Duration (V + ≤ 5.5V)	Continuous
Power Dissipation	500mW
Operating Temperature Range	
LTC144XC	0°C to 70°C
LTC144XI	40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec).	300°C

PACKAGE/ORDER INFORMATION

TOP VIEW	ORDER PART NUMBER		ORDER PART NUMBER
GND 1 8 OUT V 2 IN 3 6 REF IN 4 5 HYST	LTC1440CN8 LTC1440CS8 LTC1440IN8 LTC1440IS8	TOP VIEW GND 1 0 8 OUT V 2 0 7 V + IN 3 0 6 REF IN 4 0 5 HYST MS8 PACKAGE	LTC1440CMS8
N8 PACKAGE S8 PACKAGE 8-LEAD PDIP 8-LEAD PLASTIC SO	S8 PART MARKING	8-LEAD PLASTIC MSOP	MS8 PART MARKING
$T_{JMAX} = 150^{\circ}\text{C}, \theta_{JA} = 130^{\circ}\text{C/W} (N8)$ $T_{JMAX} = 150^{\circ}\text{C}, \theta_{JA} = 175^{\circ}\text{C/W} (S8)$	1440 1440l	T _{JMAX} = 150°C, θ _{JA} = 250°C/W	LTBX
TOP VIEW	ORDER PART NUMBER	TOP VIEW	ORDER PART NUMBER
OUT A 1 8 OUT B V 2 7 V + IN A 3 6 IN B + 5 IN B -	LTC1441CN8 LTC1441CS8 LTC1441IN8 LTC1441IS8	OUT A 1 8 OUT B V 2 7 V + IN A 3 6 REF IN B 4 5 HYST	LTC1442CN8 LTC1442CS8 LTC1442IN8 LTC1442IS8
N8 PACKAGE S8 PACKAGE 8-LEAD PDIP 8-LEAD PLASTIC SO	S8 PART MARKING	N8 PACKAGE S8 PACKAGE 8-LEAD PDIP 8-LEAD PLASTIC SO	S8 PART MARKING
$T_{JMAX} = 150$ °C, $\theta_{JA} = 130$ °C/W (N8) $T_{JMAX} = 150$ °C, $\theta_{JA} = 175$ °C/W (S8)	1441 1441I	$T_{JMAX} = 150^{\circ}C$, $\theta_{JA} = 130^{\circ}C/W$ (N8) $T_{JMAX} = 150^{\circ}C$, $\theta_{JA} = 175^{\circ}C/W$ (S8)	1442 1442I

Consult factory for Military grade parts.

ELECTRICAL CHARACTERISTICS $V^+ = 5V$, $V^- = GND = 0V$, $T_A = 25^{\circ}C$ unless otherwise noted.

SYMBO	PARAMETER	CONDITIONS				TYP	MAX	UNITS
Power S	upply							
V +	Supply Voltage Range			•	2.0		11.0	V
I _{CC}	Supply Current	IN+ = IN - = 80mV	LTC1440	•		2.1	3.7	μΑ
		HYST = REF (LTC1440/LTC1442)	LTC1441	•		3.5	5.7	μA
			LTC1442	•		3.5	5.7	μΑ



ELECTRICAL CHARACTERISTICS $V^+ = 5V$, $V^- = GND = 0V$, $T_A = 25^{\circ}C$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS				MIN	TYP	MAX	UNITS
Compara	ntor								
V _{OS}	Comparator Input Offset Voltage	V _{CM} = 2.5V			•		±3	±10	mV
I _{IN}	Input Leakage Current (IN+, IN-) Input Leakage Current (HYST)	$V_{IN}^{+} = V_{IN}^{-} = 2.5V$			•		$^{\pm 0.01}_{\pm 0.02}$	±1.0 ±1.0	nA nA
V _{CM}	Comparator Input Common Mode Range				•	٧-		V+-1.3V	V
CMRR	Common Mode Rejection Ratio	V ⁻ to V ⁺ – 1.3V					0.1	1	mV/V
PSRR	Power Supply Rejection Ratio	V+ = 2V to 11V					0.1	1	mV/V
NOISE	Voltage Noise	100Hz to 100kHz					100		μV _{RMS}
V _{HYST}	Hysteresis Input Voltage Range	LTC1440/LTC1442			•	REF – 50mV		REF	V
t _{PD}	Propagation Delay	C _{OUT} = 100pF		rive = 10mV rive = 100mV			15 8		μs μs
V _{OH}	Output High Voltage	$I_0 = -13mA$			•	V+-0.4V			V
V _{OL}	Output Low Voltage	I ₀ = 1.8mA	LTC14 LTC14	40 41/LTC1442	• •			GND + 0.4V V ⁻ + 0.4V	V V
Referenc	ce control of the con								
V_{REF}	Reference Voltage	No Load	C Tem	p Range	•	1.170	1.182	1.194	V
		LTC1440/LTC1442	I Temp	Range	•	1.164		1.200	V
		No Load LTC1440 (MSOP)	C Tem	p Range	•	1.164		1.200	V
I _{SOURCE}	Reference Output Source Current	$\Delta V_{REF} \le 1 \text{mV (LTC1442)}$			•	100	200		μΑ
I _{SINK}	Reference Output Sink Current	$\Delta V_{REF} \le 2.5 \text{mV (LTC1442)}$				10	20		μΑ
ΔV_{REF}	Reference Source Current	$0 \le I_{SOURCE} \le 2mA \text{ (LTC144)}$	10)		•		0.8	5	mV
	Reference Sink Current	$0 \le I_{SINK} \le 10\mu A \text{ (LTC1440)}$				0.5	1.5	mV	
					•			5	mV
NOISE	Voltage Noise	100Hz to 100kHz					100		μV _{RMS}
V+ = 3V	V^- = GND = 0V, T_A = 25°C unless otherwis	se noted							
Power Su		, o motou.							
V+	Supply Voltage Range				•	2		11	V
I _{CC}	Supply Current	IN+ = IN - = 80mV		LTC1440	•	_	2	3.6	 μΑ
-00	o apply o amount	HYST = REF (LTC1440/LTC	1442)	LTC1441	•		3.5	5.7	μA
		()	,	LTC1442	•		3.5	5.7	μA
Compara	itor							I.	· · ·
$\overline{V_{0S}}$	Comparator Input Offset Voltage	V _{CM} = 1.5V			•		±3	±10	mV
I _{IN}	Input Leakage Current (IN+, IN-)	$V_{IN}^{+} = V_{IN}^{-} = 1.5V$			•		±0.01	±1	nA
	Input Leakage Current (HYST)				•		±0.02	±1	nA
V _{CM}	Comparator Input Common Mode Range				•	٧-		V + - 1.3V	V
CMRR	Common Mode Rejection Ratio	V ⁻ to V ⁺ – 1.3V				0.1	1	mV/V	
PSRR	Power Supply Rejection Ratio	V+ = 2V to 11V				0.1	1	mV/V	
NOISE	Voltage Noise	100Hz to 100kHz					100		μV_{RMS}
V _{HYST}	Hysteresis Input Voltage Range	LTC1440/LTC1442			•	REF – 50mV		REF	V
t _{PD}	Propagation Delay	C _{OUT} = 100pF Overdrive = 10mV Overdrive = 100mV				14 5		μs μs	
V_{OH}	Output High Voltage	$I_0 = -8mA$		•	V+-0.4V			V	
$\overline{V_{0L}}$	Output Low Voltage	I ₀ = 0.8mA LTC1440 LTC1441/LTC1442					GND + 0.4V	V	



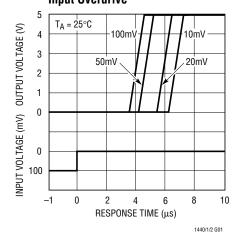
ELECTRICAL CHARACTERISTICS $V^+ = 3V$, $V^- = GND = 0V$, $T_A = 25^{\circ}C$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS			MIN	TYP	MAX	UNITS
Referenc	Reference							
$\overline{V_{REF}}$	Reference Voltage	No Load	C Temp Range	•	1.170	1.182	1.194	V
		LTC1440/LTC1442	I Temp Range	•	1.164		1.200	V
		No Load LTC1440 (MSOP)	C Temp Range	•	1.164		1.200	V
I _{SOURCE}	Reference Output Source Current	$\Delta V_{REF} \le 1 \text{mV (LTC1442)}$	$\Delta V_{REF} \le 1 \text{mV (LTC1442)}$		60	120		μА
I _{SINK}	Reference Output Sink Current	$\Delta V_{REF} \le 2.5 \text{mV (LTC1442)}$	$\Delta V_{REF} \le 2.5 \text{mV (LTC1442)}$		10	20		μА
ΔV_{REF}	Reference Source Current	0 ≤ I _{SOURCE} ≤ 1mA (LTC144	$0 \le I_{SOURCE} \le 1 \text{mA (LTC1440)}$			0.8	5.5	mV
	Reference Sink Current	0 ≤ I _{SINK} ≤ 10μA (LTC1440)				0.5	1.5	mV
				•			5	mV
NOISE	Voltage Noise	100Hz to 100kHz				100		μV _{RMS}

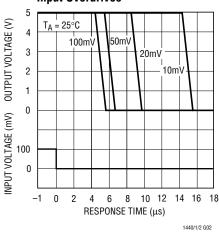
The • denotes specifications which apply over the full operating temperature range.

TYPICAL PERFORMANCE CHARACTERISTICS

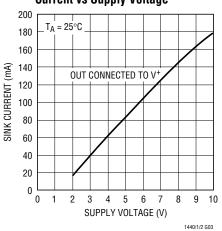
Comparator Response Time vs Input Overdrive



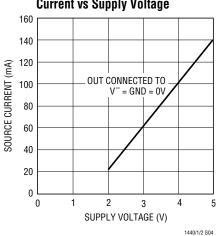
Comparator Response Time vs Input Overdrives



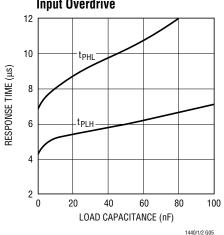
Comparator Short-Circuit Sink Current vs Supply Voltage



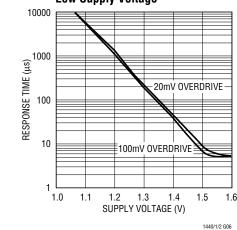
Comparator Short-Circuit Source Current vs Supply Voltage



Comparator Response Time vs Load Capacitance with 100mV Input Overdrive



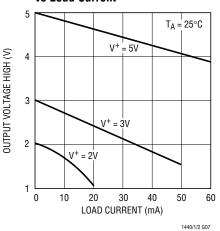
Comparator Response Time at Low Supply Voltage



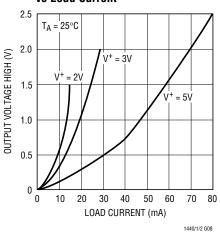


TYPICAL PERFORMANCE CHARACTERISTICS

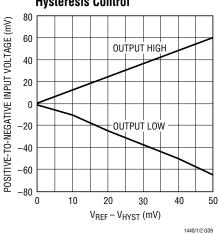
Comparator Output Voltage High vs Load Current



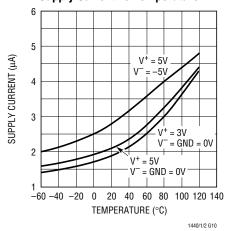
Comparator Output Voltage Low vs Load Current



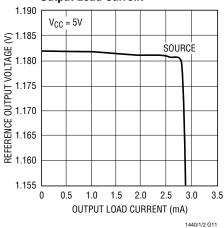
LTC1440/LTC1442 Hysteresis Control



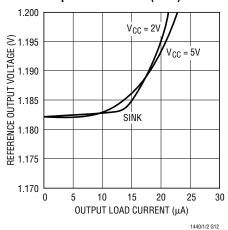
LTC1440 Supply Current vs Temperature



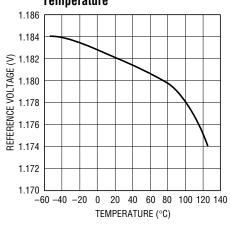
Reference Output Voltage vs Output Load Current



Reference Output Voltage vs Output Load Current (Sink)



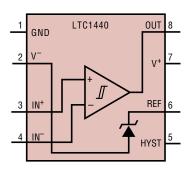
Reference Voltage vs Temperature

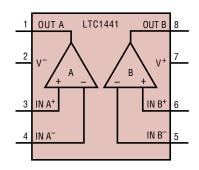


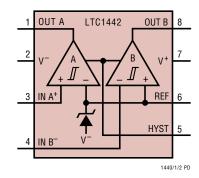
1440/1/2 G13



PIN FUNCTIONS







LTC1440

GND (Pin 1): Ground. Connect to V^- for single supply operation.

V⁻ (**Pin 2**): Negative Supply. Connect to ground for single supply operation. Potential should be more negative than GND.

IN⁺ (**Pin 3**): Noninverting Comparator Input. Input common mode range from V^- to V^+ –1.3V. Input current typically 10pA at 25°C.

IN⁻ (**Pin 4**): Inverting Comparator Input. Input common mode range from V⁻ to V⁺ –1.3V. Input current typically 10pA at 25°C.

HYST (Pin 5): Hysteresis Input. Connect to REF if not used. Input voltage range is from V_{REF} to $V_{REF} - 50 \text{mV}$.

REF (Pin 6): Reference Output. 1.182V with respect to V⁻. Can source up to $200\mu A$ and sink $15\mu A$ at $25^{\circ}C$. Drive $0.01\mu F$ bypass capacitor without oscillation.

V+ (Pin 7): Positive Supply. 2V to 11V.

OUT (Pin 8): Comparator CMOS Output. Swings from GND to V⁺. Output can source up to 40mA and sink 5mA.

LTC1441

OUT A (Pin 1): Comparator A CMOS Output. Swings from V^- to V^+ . Output can source up to 40mA and sink 5mA.

V⁻ (Pin 2): Negative Supply.

IN A⁺ (Pin 3): Noninverting Input of Comparator A. Input common mode range from V^- to V^+ –1.3V. Input current typically 10pA at 25°C.

IN A⁻ (Pin 4): Inverting Input of Comparator A. Input common mode range from V^- to V^+ –1.3V. Input current typically 10pA at 25°C.

IN B $^-$ (**Pin 5**): Inverting Input of Comparator B. Input common mode range from V $^-$ to V $^+$ -1.3V. Input current typically 10pA at 25°C.

IN B⁺ (**Pin 6**): Noninverting Input of Comparator B. Input common mode range from V^- to V^+ –1.3V. Input current typically 10pA at 25°C.

V+ (Pin 7): Positive Supply. 2V to 11V.

OUT B (Pin 8): Comparator B CMOS Output. Swings from V⁻ to V⁺. Output can source up to 40mA and sink 5mA.

LTC1442

OUT A (Pin 1): Comparator A CMOS Output. Swings from V^- to V^+ . Output can source up to 40mA and sink 5mA.

V⁻ (Pin 2): Negative Supply.

IN A⁺ (Pin 3): Noninverting Input of Comparator A. Input common mode range from V^- to V^+ –1.3V. Input current typically 10pA at 25°C.

IN B⁻ (Pin 4): Inverting Input of Comparator B. Input common mode range from V^- to V^+ –1.3V. Input current typically 10pA at 25°C.

HYST (Pin 5): Hysteresis Input. Connect to REF if not used. Input voltage range is from V_{REF} to $V_{REF} - 50$ mV.

REF (Pin 6): Reference Output. 1.182V with respect to V⁻. Can source up to $200\mu A$ and sink $15\mu A$ at $25^{\circ}C$. Drive $0.01\mu F$ bypass capacitor without oscillation.

V+ (Pin 7): Positive Supply. 2V to 11V.

OUT B (Pin 8): Comparator B CMOS Output. Swings from V⁻ to V⁺. Output can source up to 40mA and sink 5mA.

LINEAR

APPLICATIONS INFORMATION

LTC1440/LTC1441/LTC1442 are a family of micropower comparators with built-in 1.182V reference. Features include programmable hysteresis (LTC1440/LTC1442), wide supply voltage range (2V to 11V) and the ability of the reference to drive up to a $0.01\mu F$ capacitor without oscillation. The comparators' CMOS outputs can source up to 40mA and the supply current glitches, that normally occur when switching logic states, have been eliminated.

Power Supplies

The comparator family operates from a single 2V to 11V supply. The LTC1440 includes a separate ground for the comparator output stage, allowing a split supply ranging from \pm 1V to \pm 5.5V. Connecting V $^-$ to GND on the LTC1440 will allow single supply operation. If the comparator output is required to source more than 1mA, or the supply source impedance is high, V $^+$ should be bypassed with a 0.1µF capacitor.

Comparator Inputs

The comparator inputs can swing from the negative supply V^- to within 1.3V max of the positive supply V^+ . The inputs can be forced 300mV below V^- or above V^+ without damage and the typical input leakage current is only $\pm 10 pA$.

Comparator Outputs

The LTC1440 comparator output swings between GND and V+ to assure TTL compatibility with a split supply. The LTC1441 and LTC1442 outputs swing between V- and V+. The outputs are capable of sourcing up to 40mA and sinking up to 5mA while still maintaining microampere quiescent currents. The output stage does not generate crowbar switching currents during transitions which helps minimize parasitic feedback through the supply pins.

Voltage Reference

The internal bandgap reference has a voltage of 1.182V referenced to V⁻. The reference accuracy is 1.5% from -40° C to 85°C. It can source up to 200 μ A and sink up to 20 μ A with a 5V supply. The reference can drive a bypass

capacitor of up to $0.01\mu F$ without oscillation and by inserting a series resistor, capacitance values up to $100\mu F$ can be used (Figure 1).

Figure 2 shows the resistor value required for different capacitor values to achieve critical damping. Bypassing the reference can help prevent false tripping of the comparators by preventing glitches on V⁺ or reference load transients from disturbing the reference output voltage.

Figure 3 shows the bypassed reference output with a square wave applied to the V^+ pin. Resistors R2 and R3 set 10mV of hysteresis voltage band while R1 damps the reference response. Note that the comparator output doesn't trip.

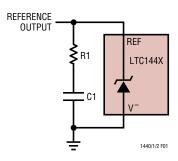


Figure 1. Damping the Reference Output

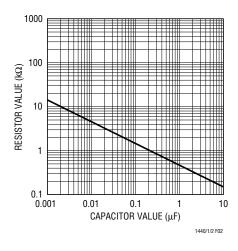


Figure 2. Damping Resistance vs Bypass Capacitor Value



APPLICATIONS INFORMATION

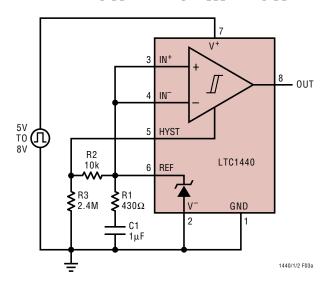


Figure 3a. Reference Transient Response Test Circuit

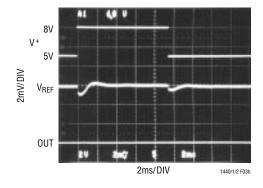


Figure 3b. Reference and Comparator Output Transient Response

Hysteresis

Hysteresis can be added to the LTC1440 by connecting a resistor (R1) between the REF and HYST pins and a second resistor (R2) from HYST to V⁻ (Figure 4).

The difference between the upper and lower threshold voltages, or hysteresis voltage band (V_{HB}), is equal to twice the voltage difference between the REF and HYST pins.

When more hysteresis is added, the upper threshold increases the same amount as the low threshold decreases. The maximum voltage allowed between REF and HYST pins is 50mV, producing a maximum hysteresis voltage band of 100mV. The hysteresis band could vary by

up to 15%. If hysteresis is not wanted, the HYST pin should be shorted to REF. Acceptable values for I_{REF} range from 0.1 μ A to 5 μ A. If 2.4M is chosen for R2, then the value of R1 is equal to the value of V_{HB} .

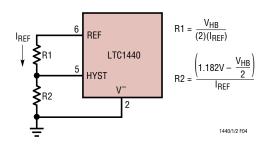


Figure 4. Programmable Hysteresis

Level Detector

The LTC1440 is ideal for use as a micropower level detector as shown in Figure 5. R1 and R2 form a voltage divider from $V_{\rm IN}$ to the noninverting comparator input. R3 and R4 set the hysteresis voltage, and R5 and C1 bypass the reference output. The following design procedure can be used to select the component values:

1. Choose the V_{IN} voltage trip level, in this example 4.65V.

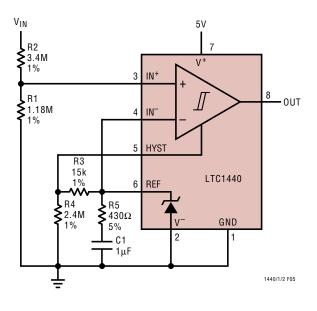


Figure 5. Glitch-Free Level Detector with Hysteresis

APPLICATIONS INFORMATION

2. Calculate the required resistive divider ratio.

Ratio =
$$V_{REF}/V_{IN}$$

Ratio = 1.182V/4.65V = 0.254

3. Choose the required hysteresis voltage band at the input V_{HBIN} , in this example 60mV. Calculate the hysteresis voltage band referred to the comparator input V_{HB} .

$$V_{HB} = (V_{HBIN})(Ratio)$$

 $V_{HB} = (60mV)(0.254)$
 $V_{HB} = 15.24mV$

4. Choose the values for R3 and R4 to set the hysteresis.

$$R4 = 2.4M$$

$$R3(k\Omega) = V_{HB} = 15k$$

5. Choose the values for R1 and R2 to set the trip point.

$$R1 = \frac{V_{REF}}{I_{BIAS}} = \frac{1.182V}{1\mu A} = 1.18M$$

$$R2 = R1 \left[\frac{V_{IN}}{V_{REF} + \frac{V_{HB}}{2}} - 1 \right]$$

$$R2 = 1.18M \left[\frac{4.65V}{1.182V + \frac{15mV}{2}} - 1 \right]$$

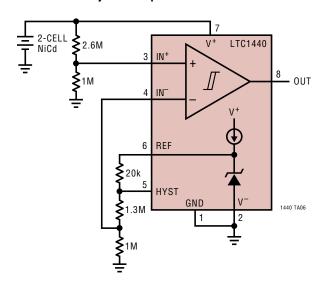
$$R2 = 3.40M$$

Low Voltage Operation

The guaranteed minimum operating voltage is 2V or $(\pm 1V)$. As the total supply voltage is reduced below 2V, the performance degrades and the supply current falls. Both the reference and the comparator are still functioning with supply voltage down to 1.5V. The reference will have slightly worse temperature coefficient. While the comparator would have less output drive and the propagation delay increases. The useful input voltage range extends from the negative supply to a little under 1V below the positive supply, which is slightly closer to the positive rail than the device operating from higher supply voltages.

TYPICAL APPLICATIONS

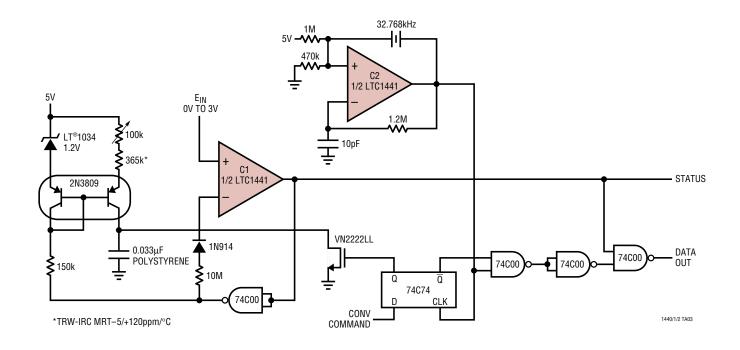
Low Battery Detect Operates Down to 1.5V



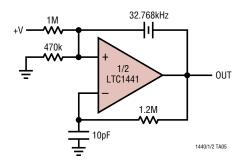


TYPICAL APPLICATIONS

10-Bit 30µA A/D Converter



32.768kHz "Watch Crystal" Oscillator

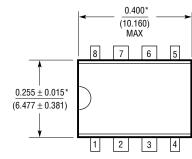


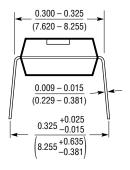
PACKAGE DESCRIPTION

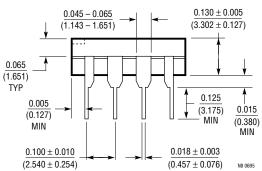
Dimensions in inches (millimeters) unless otherwise noted.

N8 Package 8-Lead PDIP (Narrow 0.300)

(LTC DWG # 05-08-1510)



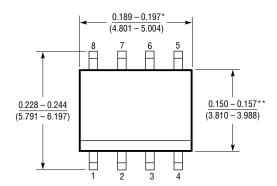


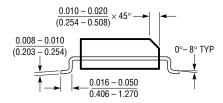


*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

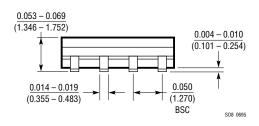
S8 Package 8-Lead Plastic Small Outline (Narrow 0.150)

(LTC DWG # 05-08-1610)





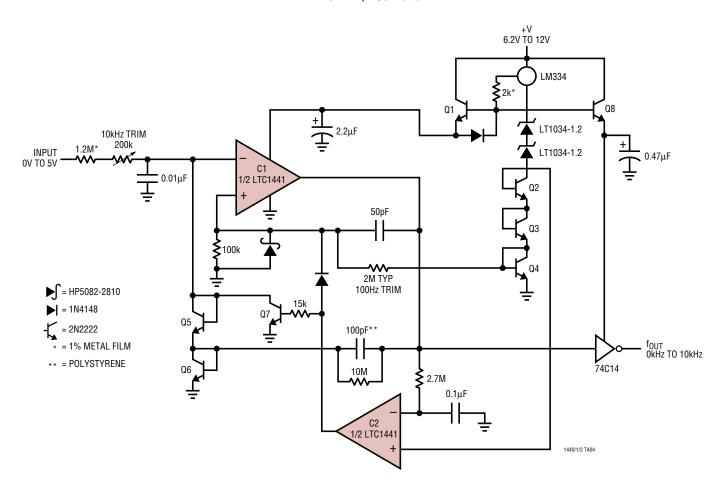
- * DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE
- **DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE





TYPICAL APPLICATION

10kHz V/F Converter



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1178/LT1179	Dual/Quad 17μA Precision Single Supply Op Amps	70μV Max V _{OS} , 5nA Max I _{BIAS}
LT1351	Single 250µA, 3MHz, 200V/µs Op Amp with Shutdown	C-Load [™] Op Amp Stable Driving Any Capacitive Load
LT1352/LT1353	Dual/Quad 250µA, 3MHz, 200V/µs Op Amps	C-Load Op Amps Stable Driving Any Capacitive Load
LTC1443/LTC1444/LTC1445	Micropower Quad Comparator with 1% Reference	LTC1443 Has 1.182V Reference, LTC1444/LTC1445 Have 1.221V Reference and Adjustable Hysteresis

C-Load is a trademark of Linear Technology Corporation.

